

3315\_Ch4\_6\_Ch6\_Presentation3

**FREQUENCY RESPONSE OF FIR FILTERS**

**RECORD ON**

# LECTURE OBJECTIVES

- **SINUSOIDAL** INPUT SIGNAL
  - DETERMINE the FIR FILTER OUTPUT
- **FREQUENCY RESPONSE** of FIR
  - PLOTTING vs. Frequency
  - MAGNITUDE vs. Freq
  - PHASE vs. Freq

$$H(e^{j\hat{\omega}}) = |H(e^{j\hat{\omega}})| e^{j\angle H(e^{j\hat{\omega}})}$$

Amplitude and magnitude , both these terms are similar with a minute difference.

The amplitude of a variable is the measure of how far, and in what direction, that variable differs from zero. Thus, signal amplitudes can be either positive or negative.

The **magnitude** of a variable, on the other hand, is the measure of how far, regardless of direction, its quantity differs from zero. So magnitudes are always positive values.

Magnitude of  $x$  is represented by  $|x|$  symbol. In 1D this is Absolute value. In 2D, it is length of a vector

SOURCE :

UNDERSTANDING DSP by RICHARD G. LYONS

**1. INPUT**  $A \cos(2\pi (100)t)$  TO A LINEAR SYSTEM

**OUTPUT**  $B \cos(2\pi (100)t + \varphi)$

**CHANGE IN AMPLITUDE** from A to B

**AND PHASE** from **0** to  $\varphi$  (TYPICALLY A TIME DELAY)

$$2. Ae^{j\omega} = A \cos(2\pi f) + jA \sin(2\pi f) = |A| \angle \omega$$

THIS IS A VECTOR WITH **MAGNITUDE** (LENGTH) A  
AND **ANGLE IN**  $\omega$  **2D**

# TIME & FREQUENCY

$$y[n] = \sum_{k=0}^M b_k x[n-k] = \sum_{k=0}^M h[k] x[n-k]$$

FIR DIFFERENCE EQUATION is the TIME-DOMAIN

$$H(e^{j\hat{\omega}}) = \sum_{k=0}^M h[k] e^{-j\hat{\omega}k}$$

$$\begin{aligned} H(e^{j\hat{\omega}}) &= h[0] + h[1]e^{-j\hat{\omega}} + h[2]e^{-j2\hat{\omega}} + h[3]e^{-j3\hat{\omega}} + \dots \\ &= |H(e^{j\hat{\omega}})| e^{j\angle H(e^{j\hat{\omega}})} \end{aligned}$$

# COMPLEX EXPONENTIAL

$$x[n] = Ae^{j\varphi} e^{j\hat{\omega}n} \quad -\infty < n < \infty$$

**x[n] is the input signal—a complex exponential**

$$y[n] = \sum_{k=0}^M b_k x[n-k] = \sum_{k=0}^M h[k] x[n-k]$$

**FIR DIFFERENCE EQUATION**

# COMPLEX EXP OUTPUT

- Use the FIR “Difference Equation”

$$y[n] = \sum_{k=0}^M b_k x[n-k] = \sum_{k=0}^M b_k A e^{j\varphi} e^{j\hat{\omega}(n-k)}$$

$$= \left( \sum_{k=0}^M b_k e^{j\hat{\omega}(-k)} \right) A e^{j\varphi} e^{j\hat{\omega}n}$$

$$= H(\hat{\omega}) A e^{j\varphi} e^{j\hat{\omega}n}$$

**H** IS THE TRANSFER FUNCTION

# FREQUENCY RESPONSE

EXAMPLE 6.1  $b_k = \{1, 2, 1\}$  so

$$H(e^{j\hat{\omega}}) = 1 + 2e^{-j\hat{\omega}} + e^{-j2\hat{\omega}}$$

write as  $= e^{-j\hat{\omega}} (2 + 2\cos\hat{\omega})$  to plot

$$-\pi < \hat{\omega} < \pi$$

This is  $|H(e^{j\hat{\omega}})| \frac{1}{e^{-j\hat{\omega}}}$





# EXAMPLE 6.1

REMEMBER  $-\pi < \hat{\omega} < \pi$

$$\{b_k\} = \{1, 2, 1\}$$

$$\begin{aligned} H(e^{j\hat{\omega}}) &= 1 + 2e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} \\ &= e^{-j\hat{\omega}} (e^{j\hat{\omega}} + 2 + e^{-j\hat{\omega}}) \\ &= e^{-j\hat{\omega}} (2 + 2\cos \hat{\omega}) \end{aligned}$$

EXPLOIT  
SYMMETRY

Since  $(2 + 2\cos \hat{\omega}) \geq 0$

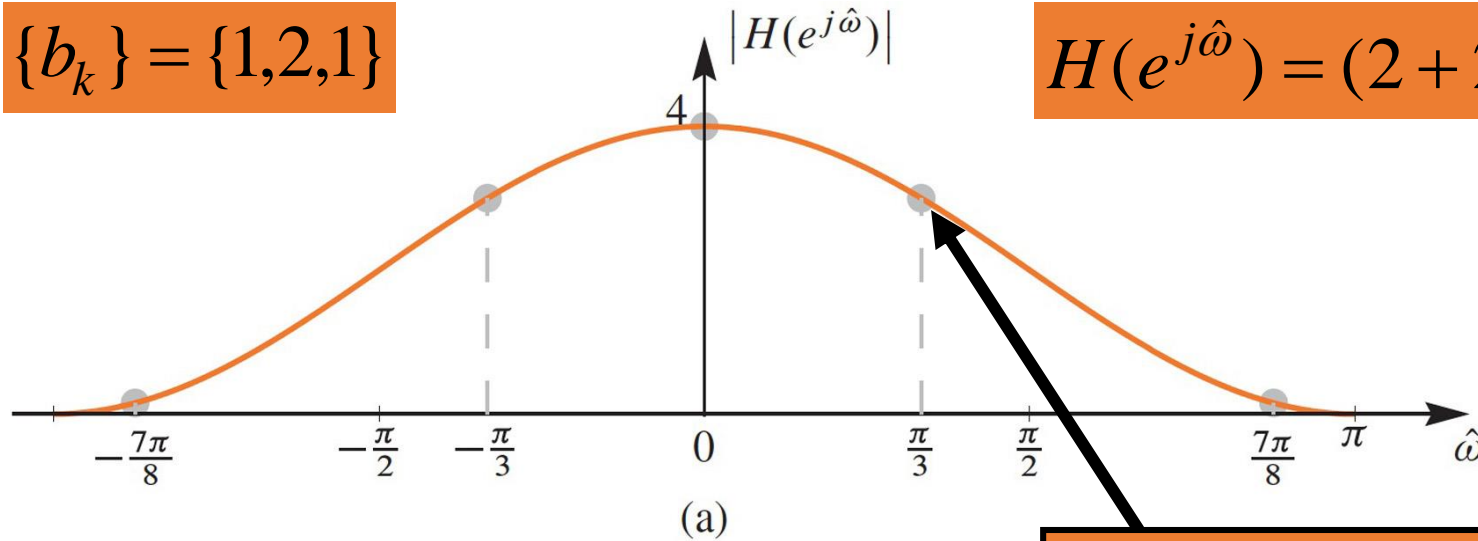
Magnitude is  $|H(e^{j\hat{\omega}})| = (2 + 2\cos \hat{\omega})$

and Phase is  $\angle H(e^{j\hat{\omega}}) = -\hat{\omega}$

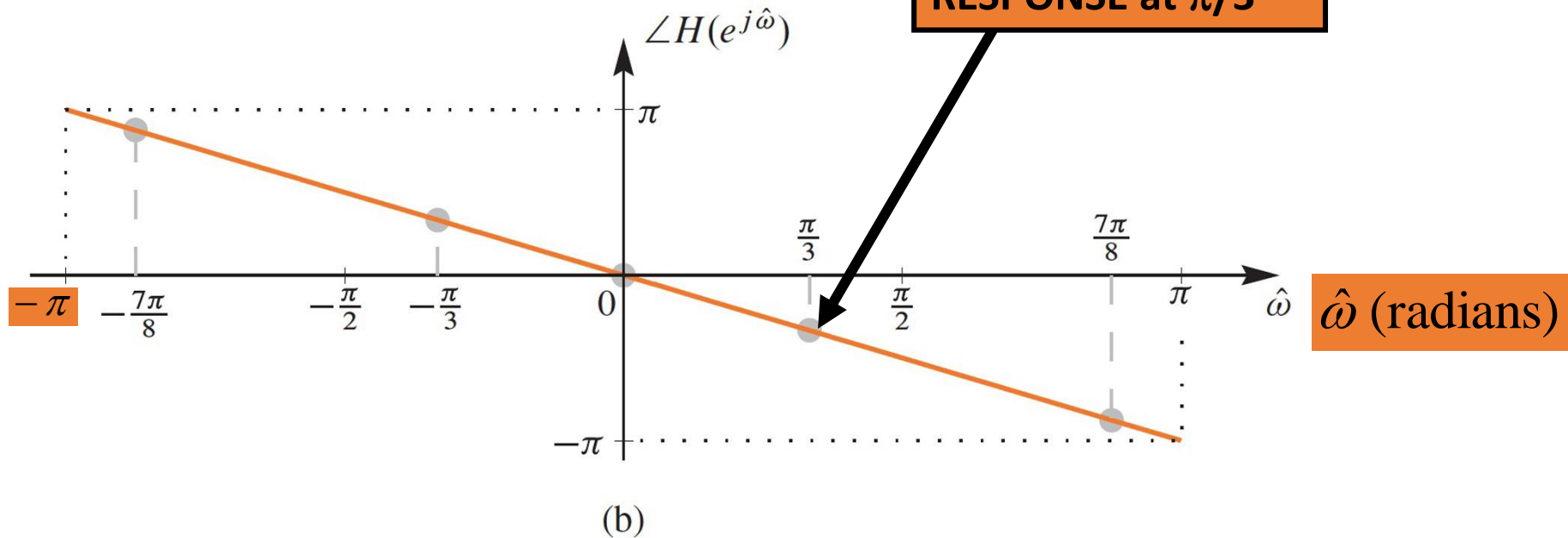
# PLOT of FREQ RESPONSE

$$\{b_k\} = \{1, 2, 1\}$$

$$H(e^{j\hat{\omega}}) = (2 + 2\cos\hat{\omega})e^{-j\hat{\omega}}$$



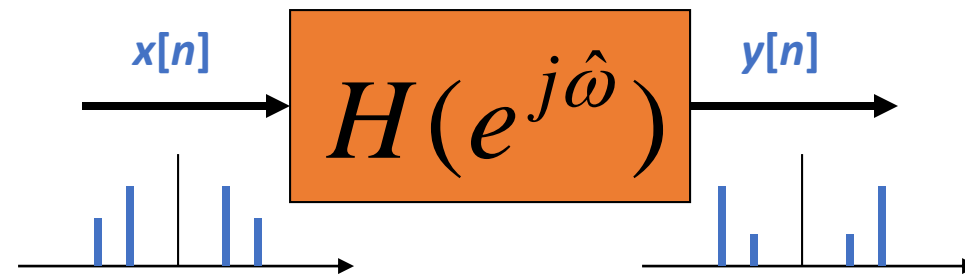
**RESPONSE at  $\pi/3$**



$\hat{\omega}$  (radians)

## EXAMPLE 6.2

Find  $y[n]$  when  $H(e^{j\hat{\omega}})$  is known  
and  $x[n] = 2e^{j\pi/4} e^{j(\pi/3)n}$



$$H(e^{j\hat{\omega}}) = (2 + 2 \cos \hat{\omega}) e^{-j\hat{\omega}}$$

## EXAMPLE 6.2 (answer)

Find  $y[n]$  when  $x[n] = 2e^{j\pi/4} e^{j(\pi/3)n}$

Evaluate  $H(e^{j\hat{\omega}})$  at  $\hat{\omega} = \pi/3$

$$H(e^{j\hat{\omega}}) = (2 + 2\cos\hat{\omega})e^{-j\hat{\omega}}$$

$$H(e^{j\hat{\omega}}) = 3e^{-j\pi/3} \quad @ \hat{\omega} = \pi/3$$

$$y[n] = \left(3e^{-j\pi/3}\right) \times 2e^{j\pi/4} e^{j(\pi/3)n} = 6e^{-j\pi/12} e^{j(\pi/3)n}$$

### EXAMPLE 6-3 Cosine Input

For the FIR filter with coefficients  $\{b_k\} = \{1, 2, 1\}$ , find the output when the input is

$$x[n] = 3 \cos\left(\frac{\pi}{3}n - \frac{\pi}{2}\right)$$

The frequency response of the system was determined in Example 6-1 to be

$$H(e^{j\hat{\omega}}) = (2 + 2 \cos \hat{\omega})e^{-j\hat{\omega}}$$

Note that  $H(e^{-j\hat{\omega}}) = H^*(e^{j\hat{\omega}})$ , so  $H(e^{j\hat{\omega}})$  has conjugate symmetry. Solution of this problem requires just one evaluation of  $H(e^{j\hat{\omega}})$  at the frequency  $\hat{\omega} = \pi/3$ :

$$\begin{aligned} H(e^{j\pi/3}) &= e^{-j\pi/3} (2 + 2 \cos(\pi/3)) \\ &= e^{-j\pi/3} (2 + 2(\frac{1}{2})) = 3e^{-j\pi/3} \end{aligned}$$

Therefore, the magnitude is  $|H(e^{j\pi/3})| = 3$  and the phase is  $\angle H(e^{j\pi/3}) = -\pi/3$ , so the output is

$$\begin{aligned} y[n] &= (3)(3) \cos\left(\frac{\pi}{3}n - \frac{\pi}{3} - \frac{\pi}{2}\right) \\ &= 9 \cos\left(\frac{\pi}{3}(n-1) - \frac{\pi}{2}\right) \end{aligned}$$

Notice that the magnitude of the frequency response multiplies the amplitude of the cosine signal, and the phase angle of the frequency response adds to the phase of the signal. This problem could also be studied by using the discrete-time Fourier transform.

## Example: Lowpass Averager

- Consider a 5-point moving average filter wrapped up between a C-to-D and D-to-C system
- We assume a sampling rate of 1000 Hz and an input composed of two sinusoids

$$x(t) = \cos[2\pi(100)t] + 3\cos[2\pi(300)t]$$

- Find the system frequency response in terms of the analog frequency variable  $f$ , and find the steady-state output  $y(t)$
- We will use `freqz()` to obtain the frequency response

```
>> w = -pi:pi/100:pi;  
>> H = freqz(ones(1,5)/5,1,w);  
>> subplot(211)
```

SOLVE THE PROBLEM BEFORE MATLAB – IF POSSIBLE!  
AT LEAST KNOW THE RANGES INVOLVED

$$f_s = 1000 \text{ Hz} \quad \text{so} \quad f_{\max} = 500 \text{ Hz}$$

$$x(t) = \cos\{2\pi(100t)\} + 3\cos\{2\pi(300t)\}$$

$$\hat{\omega} = \omega T_s = \frac{\omega}{f_s} \quad \hat{\omega}_{100} = 2\pi \frac{100}{1000} = 0.2\pi \quad \hat{\omega}_{300} = 0.6\pi$$

$$\text{Range of } \hat{\omega} \quad -\frac{500}{1000} \times 2\pi < \hat{\omega} < \frac{500}{1000} 2\pi$$

$$\begin{array}{l} \text{Digital PLOT} \quad -\pi < \hat{\omega} < \pi \\ \text{Analog PLOT} \quad -500 \text{ Hz} < f < 500 \text{ Hz} \end{array}$$

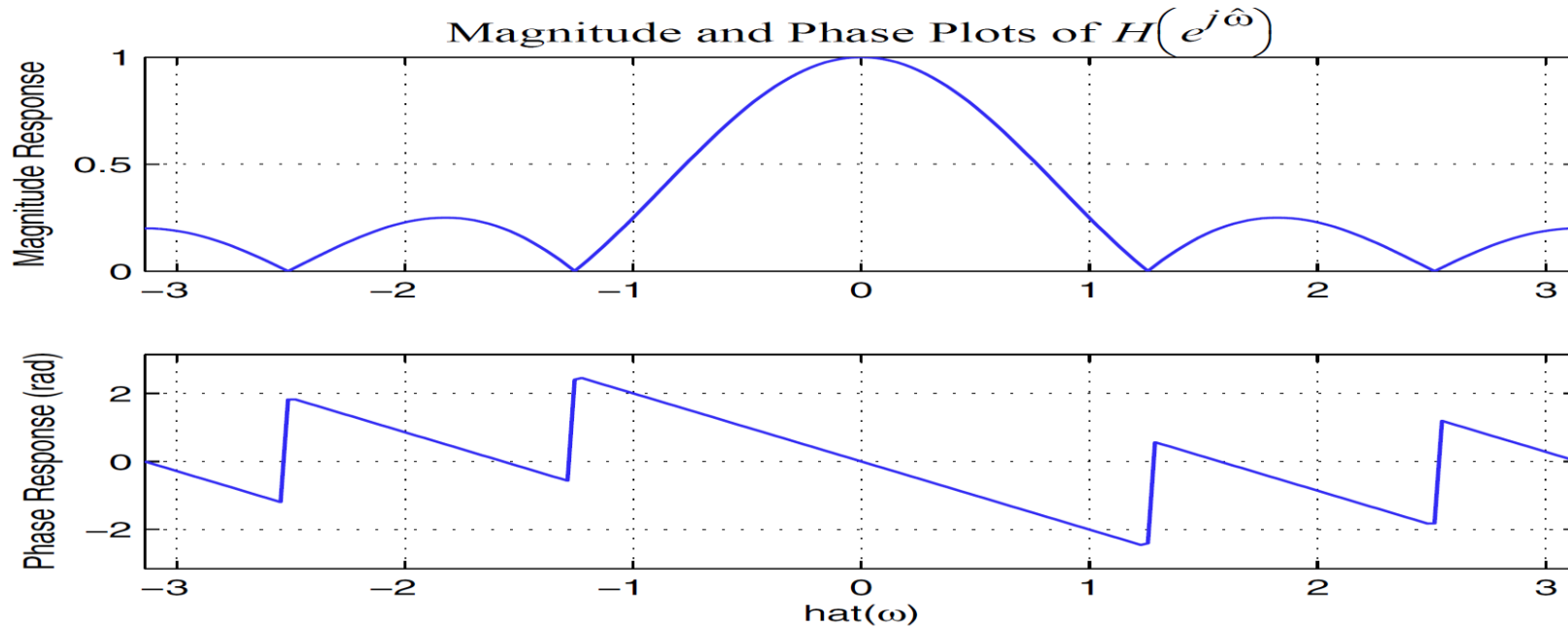


```

>> plot(w,abs(H))
>> axis([-pi pi 0 1]); grid
>> ylabel('Magnitude Response')
>> subplot(212)
>> plot(w,angle(H))
>> axis([-pi pi -pi pi]); grid
>> ylabel('Phase Response (rad)')
>> xlabel('hat(\omega)')

```

Range  $-\pi$  to  $\pi$



```

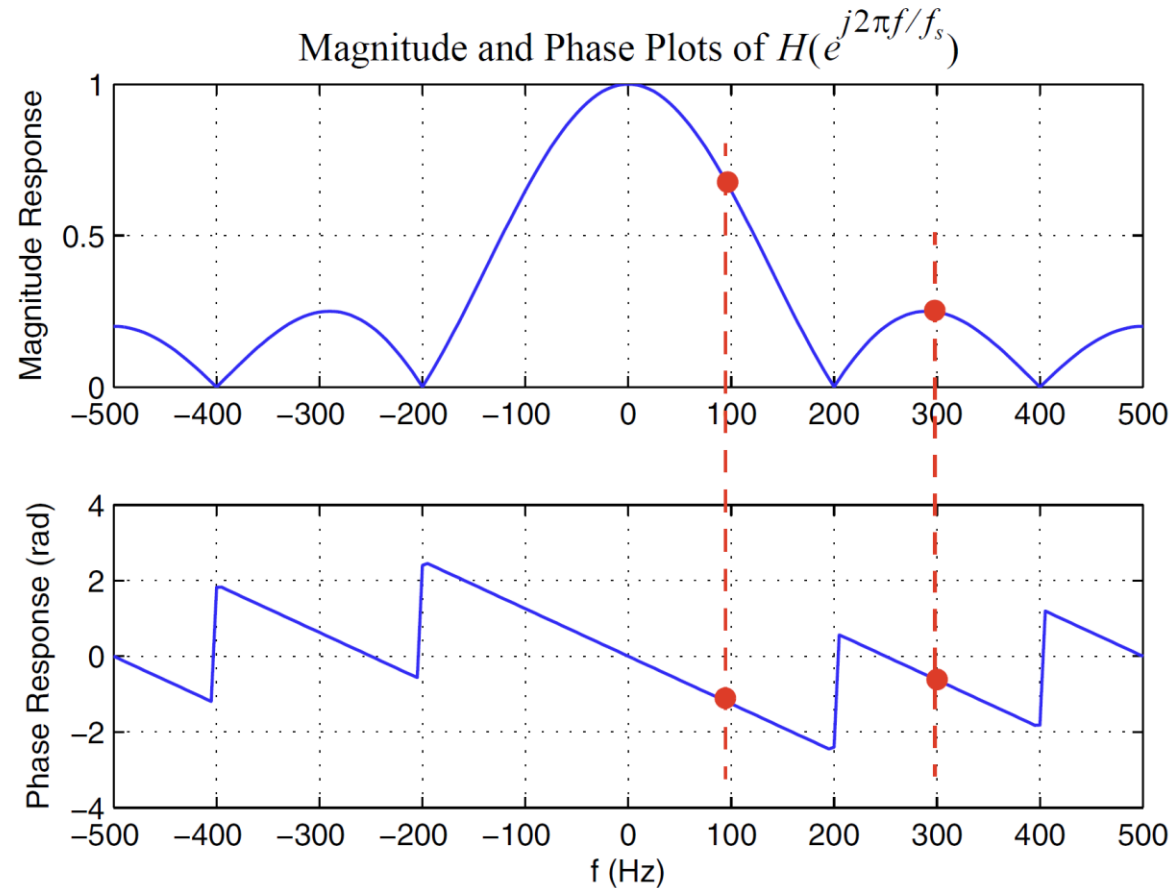
>> subplot(211)
>> plot(w*1000/(2*pi),abs(H))
>> grid
>> ylabel('Magnitude Response')
>> subplot(212)
>> plot(w*1000/(2*pi),angle(H))
>> grid
>> ylabel('Phase Response (rad)')
>> xlabel('f (Hz)')

```

**IN THE DIGITAL FREQUENCY DOMAIN**

# IN THE ANALOG FREQUENCY DOMAIN

*Filtering Sampled Continuous-Time Signals*



EACH POINT IS A  
TRANSFER VALUE  
SHOWING THE CHANGE  
IN MAGNATUDE AND  
PHASE AT EACH  
FREQUENCY.

DO A GOOD JOB ON QUIZ = HW5

RECORD OFF