

MATLAB COMMANDS

DSP USES



conv Convolution and polynomial multiplication.

$C = \text{conv}(A, B)$ convolves vectors A and B.

Syntax

$w = \text{conv}(u, v)$

Description

[example](#)

$w = \text{conv}(\underline{u}, \underline{v})$ returns the [convolution](#) of vectors u and v.

If u and v are vectors of polynomial coefficients, convolving them is equivalent to multiplying the two polynomials.

```
% Create vectors u and v containing the coefficients of the  
% polynomials  $X^2 + 1$  and  $2X + 7$  .
```

```
u = [1 0 1]; v = [2 7]; % Coefficients
```

Use convolution to multiply the polynomials.

```
w = conv(u,v)
```

```
w = 2 7 2 7 % 3th order Product
```

w contains the polynomial coefficients for
 $2X^3 + 7X^2 + 2X + 7$

```
u = [1 1 1];  
v = [1 1 0 0 0 1 1];  
w = conv(u,v)
```

1 2 2 1 0 1 2 2 1

The length of w is $\text{length}(u) + \text{length}(v) - 1$,
which in this example is 9.

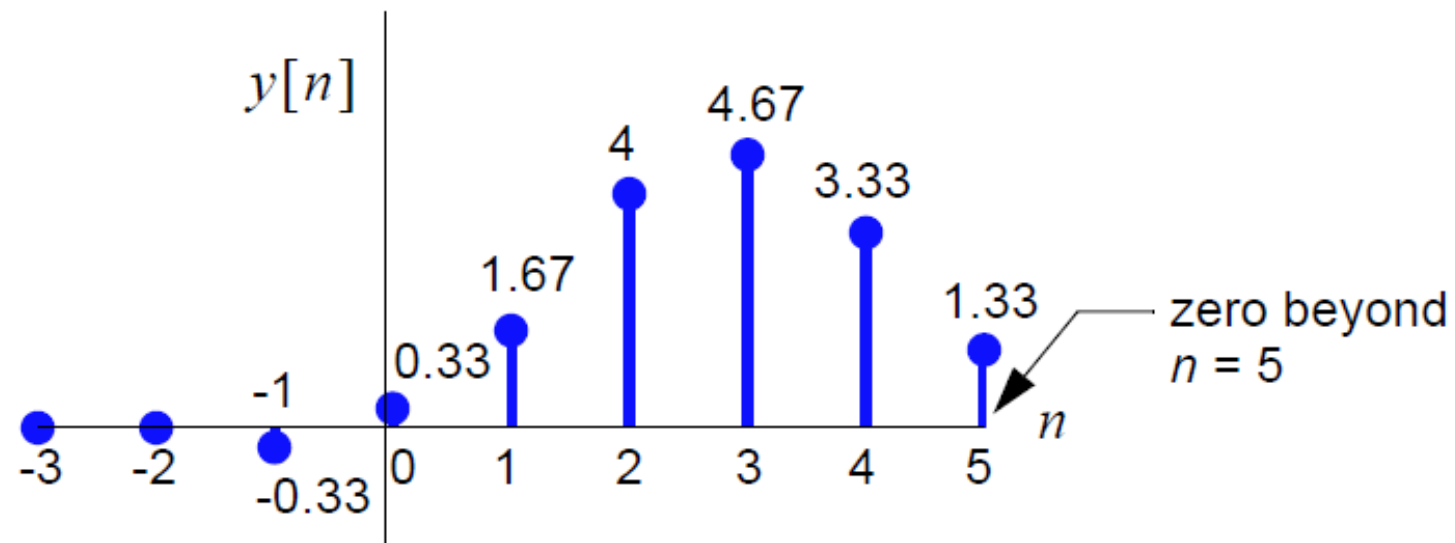
filter One-dimensional digital filter.

$Y = \text{filter}(B,A,X)$ filters the data in vector X with the filter described by vectors A and B to create the filtered data Y . The filter is a "Direct Form II Transposed" implementation of the standard difference equation:

$$a(1)*y(n) = b(1)*x(n) + b(2)*x(n-1) + \dots + b(nb+1)*x(n-nb) \\ - a(2)*y(n-1) - \dots - a(na+1)*y(n-na)$$

If $a(1)$ is not equal to 1, filter normalizes the filter coefficients by $a(1)$.

```
>> n= -3:5;  
>> x = [0 0 -1 2 4 6 4 0 0]  
>> % We will learn about the filter function later  
>> y = filter(1/3*[1 1 1],1,x);  
>> stem(n,y,'filled')
```



$a = 1, b_k = 1/3 * [1 \ 1 \ 1]$

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Example: $\{b_k\} = \{1, 1, 3, 1, 1\}$

- The frequency response of this FIR filter is

$$\begin{aligned} H(e^{j\hat{\omega}}) &= \sum_{k=0}^4 b_k e^{-j\hat{\omega}k} \\ &= 1 + e^{-j\hat{\omega}} + 3e^{-j2\hat{\omega}} + e^{-j3\hat{\omega}} + e^{-j4\hat{\omega}} \end{aligned}$$

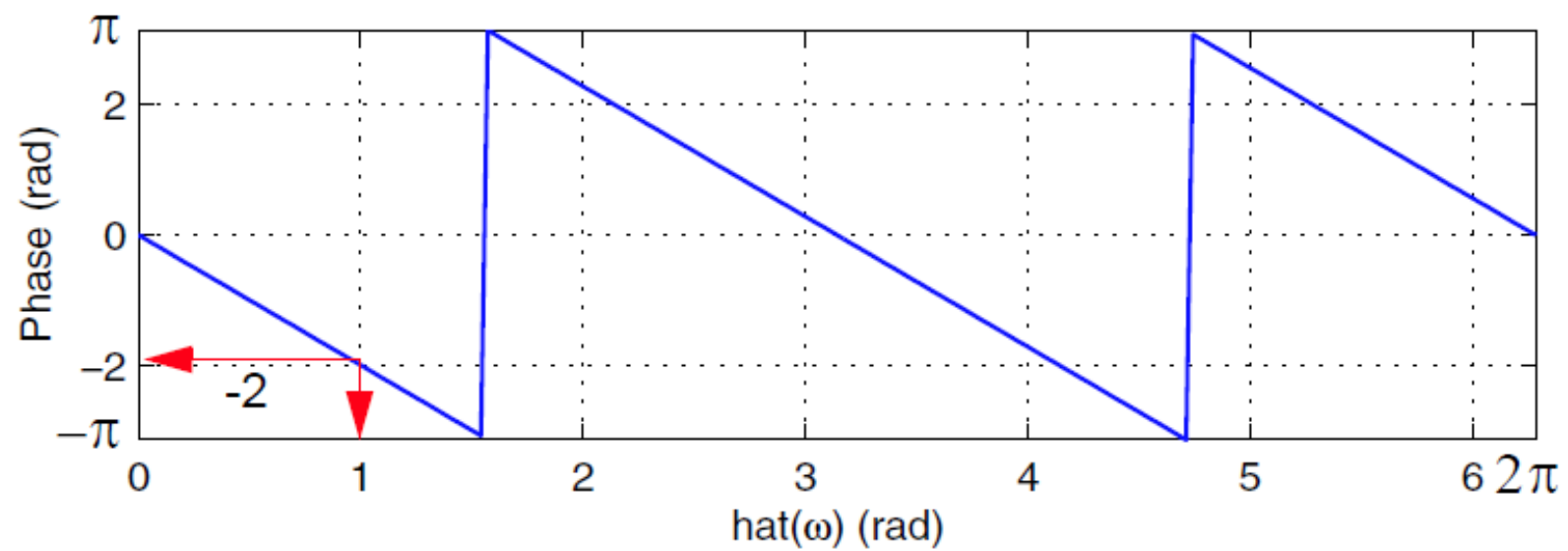
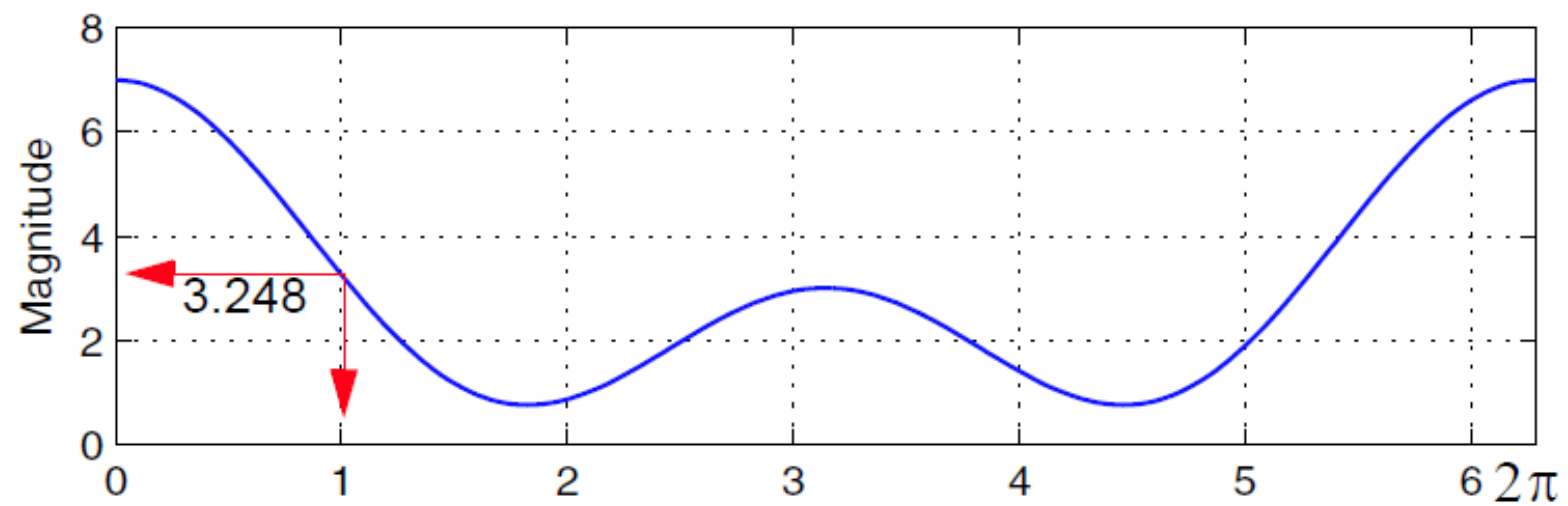
$$\begin{aligned}
 H(e^{j\hat{\omega}}) &= e^{-j2\hat{\omega}} [e^{j2\hat{\omega}} + e^{j\hat{\omega}} + 3 + e^{-j\hat{\omega}} + e^{-j2\hat{\omega}}] \\
 &= e^{-j2\hat{\omega}} [2\cos(2\hat{\omega}) + 2\cos(\hat{\omega}) + 3]
 \end{aligned}$$

- We have used the inverse Euler formula for cosine twice
- For this particular filter we have that

$$\begin{aligned}
 |H(e^{j\hat{\omega}})| &= 3 + 2\cos(\hat{\omega}) + 2\cos(2\hat{\omega}) \\
 \angle H(e^{j\hat{\omega}}) &= -2\hat{\omega}
 \end{aligned}$$

- Use MATLAB to plot the magnitude and phase response

```
>> w = 0:2*pi/200:2*pi;  
>> H = exp(-j*2*w) .* (3 + 2*cos(w) + 2*cos(2*w));  
>> subplot(211)  
>> plot(w,abs(H))  
>> axis([0 2*pi 0 8])  
>> grid  
>> ylabel('Magnitude')  
>> subplot(212)  
>> plot(w,angle(H))  
>> axis([0 2*pi -pi pi])  
>> grid  
>> ylabel('Phase (rad)')  
>> xlabel('hat(\omega) (rad)')
```



Example: Find $y[n]$ for Input $x[n] = 5e^{j(1 \cdot n)}$

- The input frequency is $\hat{\omega}_0 = 1$ rad, the amplitude is 5, and the phase is $\phi = 0$
- Assuming $x[n]$ is input to the 4-tap FIR filter in the previous example, the filter output is

$$\begin{aligned}y[n] &= (3 + 2\cos(1) + 2\cos(2 \cdot 1))e^{-j(1 \cdot 2)}5e^{j(1 \cdot n)} \\ &= 16.2415e^{-j2} \cdot e^{jn}\end{aligned}$$

- The amplitude response or gain at $\hat{\omega}_0 = 1$ is $|H(e^{j1})| = 3.248$; why?

- For this particular filter we have that

$$\left| H(e^{j\hat{\omega}}) \right| = 3 + 2\cos(\hat{\omega}) + 2\cos(2\hat{\omega})$$

Approximate this at theta = 1 rad; $\cos(1)$ about $\cos(60) \approx \frac{1}{2}$
 $\cos(2)$ about $\cos(120) \approx -\frac{1}{2}$.

Expect $3 + 2 \cdot .5 + 2(-0.5) = \text{about } 3$.

Exact $3 + 1.0806 - 0.8323 = 3.2843$

Actually $1/\pi = \text{theta}/180 = 1/(3.14) \cdot 180 = 1(3 + .14/3) \cdot 180 = 60 - 60(.05)$ ABOUT 57 DEGREES
 $1/(1+X) \approx 1-X$ from Taylor $X = 0.14/3 \approx 0.05$

Example: Three Inputs with $b_k = \{1, -1, 1\}$

- The input is

$$x[n] = 10 + 4 \cos\left(\frac{\pi}{4}n + \frac{\pi}{8}\right) + 3 \cos\left(\frac{\pi}{3}n - \frac{\pi}{4}\right)$$

- The frequency response is

$$\begin{aligned} H(e^{j\hat{\omega}}) &= 1 - e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} \\ &= e^{-j\hat{\omega}}[-1 + 2 \cos(\hat{\omega})] \end{aligned}$$

- The input frequencies are $\hat{\omega}_k = \{0, \pi/4, \pi/3\}$

$$H(e^{j0}) = e^{-j0}[-1 + 2 \cos(0)] = 1$$

$$H(e^{j\pi/4}) = e^{-j\pi/4}[-1 + 2 \cos(\pi/4)] = 0.4142 e^{-j\pi/4}$$

$$H(e^{j\pi/3}) = e^{-j\pi/3}[-1 + 2 \cos(\pi/3)] = 0$$

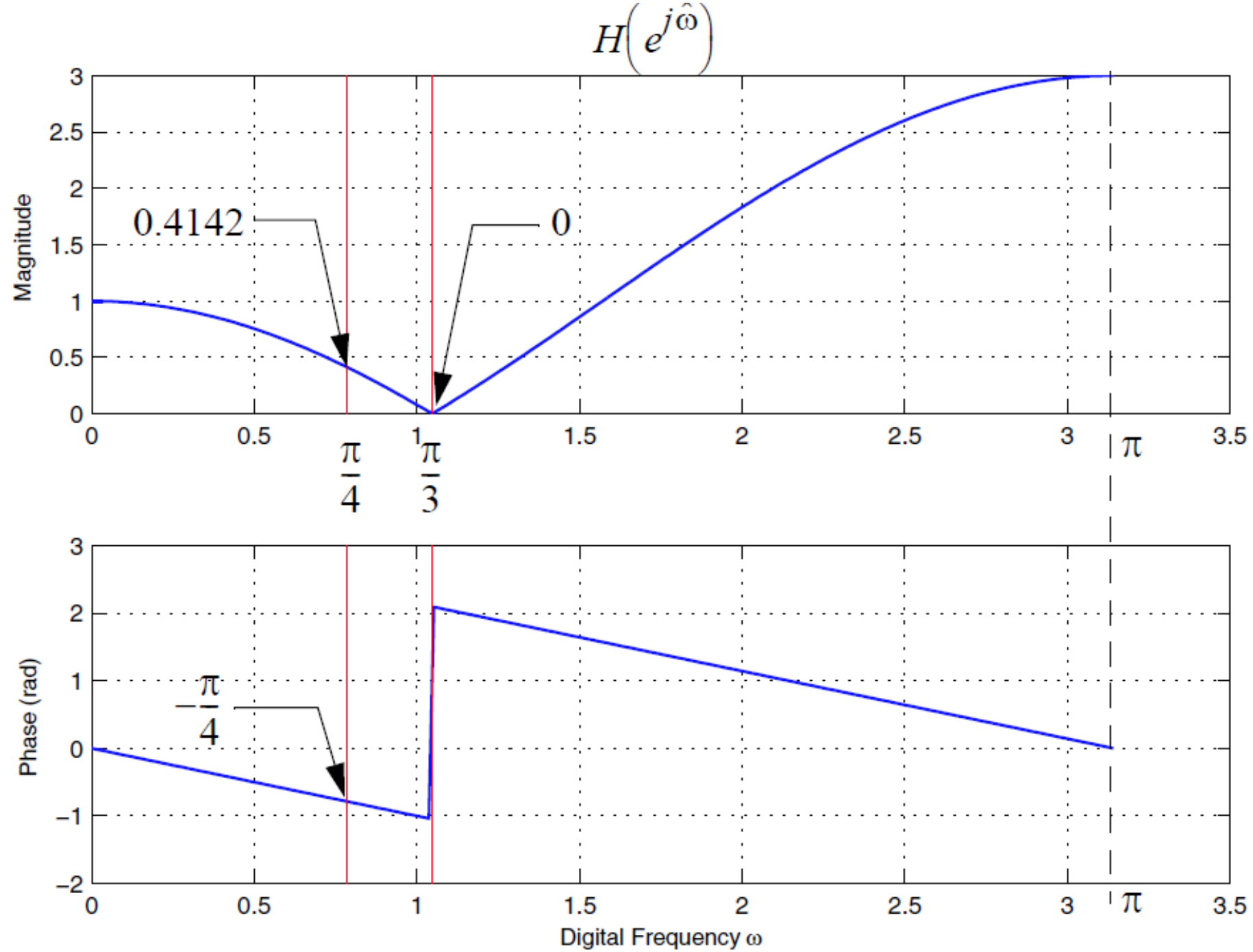
- Thus

$$\begin{aligned} y[n] &= 10 \cdot 1 + 4 \cdot 0.4142 \cos\left(\frac{\pi}{4}n + \frac{\pi}{8} - \frac{\pi}{4}\right) \\ &= 10 + 1.6569 \cos\left(\frac{\pi}{4}n - \frac{\pi}{8}\right) \end{aligned}$$

```

>> w = 0:pi/200:pi;
>> H = 1 - exp(-j*w) + exp(-j*2*w);
>> subplot(211)
>> plot(w,abs(H))
>> grid
>> hold
>> plot([pi/4 pi/4],[0 3],'r')
>> plot([pi/3 pi/3],[0 3],'r')
>> ylabel('Magnitude')
>> subplot(212)
>> plot(w,angle(H))
>> grid
>> hold
>> plot([pi/4 pi/4],[-2 3],'r')
>> ylabel('Phase (rad)')
>> xlabel('Digital Frequency \omega')

```



```
>> help freqz
```

FREQZ Digital filter frequency response.

[H,W] = FREQZ(B,A,N) returns the N-point complex frequency response vector H and the N-point frequency vector W in radians/sample of the filter:

$$H(e^{j\omega}) = \frac{B(e^{j\omega})}{A(e^{j\omega})} = \frac{b(1) + b(2)e^{-j\omega} + \dots + b(m+1)e^{-jm\omega}}{a(1) + a(2)e^{-j\omega} + \dots + a(n+1)e^{-jn\omega}}$$

given numerator and denominator coefficients in vectors B and A. The frequency response is evaluated at N points equally spaced around the upper half of the unit circle. If N isn't specified, it defaults to 512.

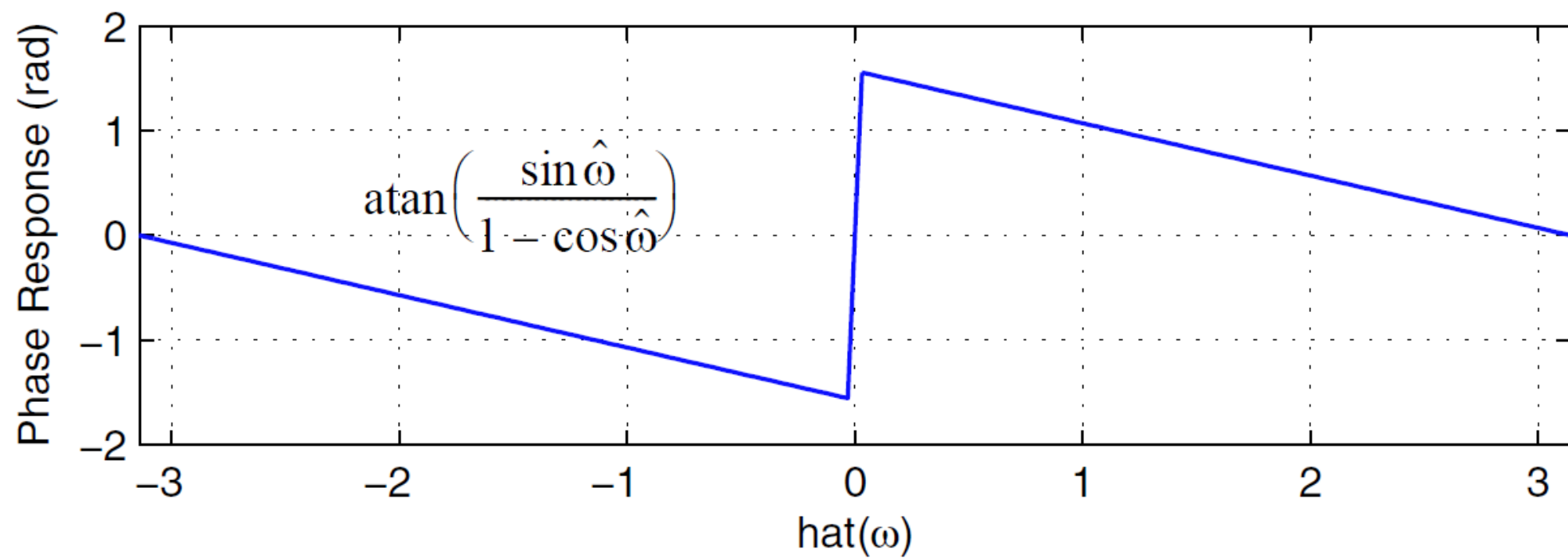
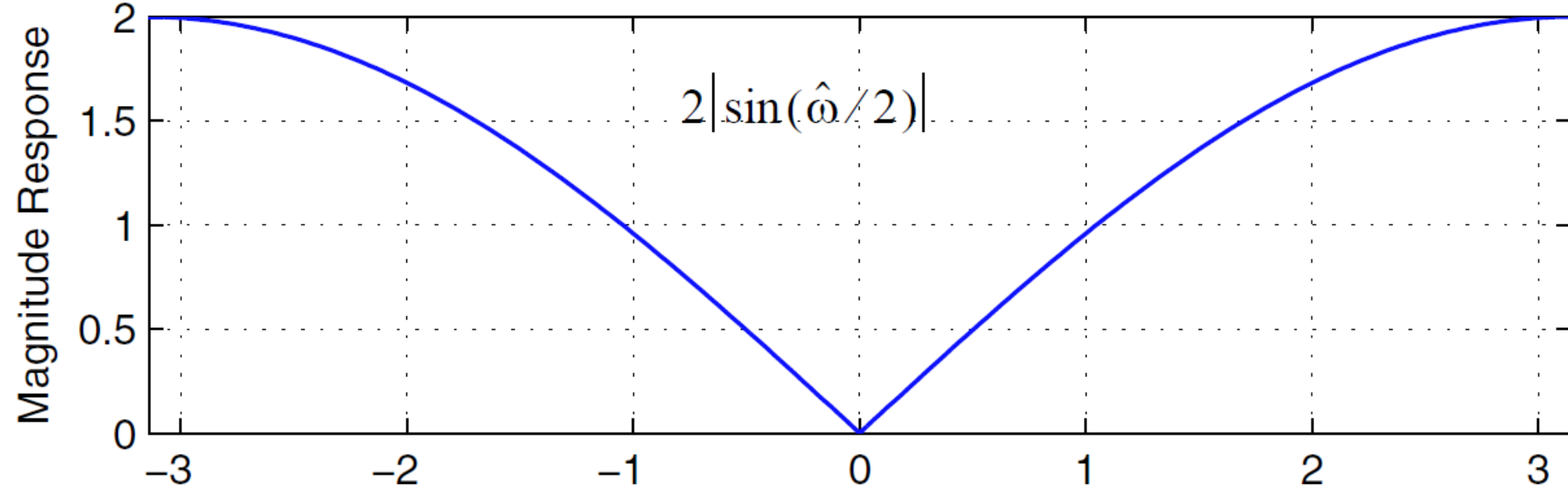
Example: First-Difference System $y[n] = x[n] - x[n - 1]$

- The frequency response is

$$H(e^{j\hat{\omega}}) = 1 - e^{-j\hat{\omega}} = (1 - \cos \hat{\omega}) + j \sin \hat{\omega}$$

```
>> w = -pi:pi/100:pi;  
>> H = freqz([1 -1],1,w); %use a custom w axis  
>> subplot(211)  
>> plot(w,abs(H))  
>> axis([-pi pi 0 2]); grid; ylabel('Magnitude Response')  
>> subplot(212)  
>> plot(w,angle(H))  
>> axis([-pi pi -2 2]); grid;  
>> ylabel('Phase Response (rad)')  
>> xlabel('hat(\omega)')
```

KILL THE dc component



```
>> help diric
```

diric Dirichlet, or periodic sinc function

Y = diric(X,N) returns a matrix the same size as X whose elements are the Dirichlet function of the elements of X. Positive integer N is the number of equally spaced extrema of the function in the interval 0 to 2*pi.

The Dirichlet function is defined as

$$d(x) = \sin(Nx/2) ./ (N \sin(x/2)) \quad \text{for } x \text{ not a multiple of } 2\pi$$
$$+1 \text{ or } -1 \text{ for } x \text{ a multiple of } 2\pi. \text{ (depending on limit)}$$

% Example 1:

% Plot the Dirichlet function over the range 0 to 4, for N = 7 and % N = 8.

```
x = linspace(0,4*pi,300);  
subplot(211); plot(x,diric(x,7));  
title('Diric, N = 7'); axis tight;  
subplot(212); plot(x,diric(x,8));  
title('Diric, N = 8'); axis tight;
```

