Harman Outline 1B CENG 5131 PDF

D. Polynomials (Harman P 160)

Review Harman Pages 160-161.

Let a_0, a_1, \ldots, a_n be n+1 arbitrary numbers with $a_n \neq 0$. Then, the function

$$P(z) = a_n z^n + a_{n-1} z^{n-1} + \dots + a_0 \tag{1}$$

is a *polynomial* of degree n. The n+1 constants a_0, a_1, \ldots, a_n are the *coefficients* of the polynomial. A polynomial is a *real polynomial* if all its coefficients are real numbers. This text considers only polynomials with real coefficients unless otherwise stated, because these are associated with mathematical models of physical systems.

The numbers z that are solutions to the equation

$$P(z) = 0 \tag{2}$$

are called the *roots* or sometimes the *zeros* of the polynomial. The values of the roots are not necessarily real numbers. Thus, a root z may have the form z = x + iy, where i is the imaginary number $\sqrt{-1}$. In electrical engineering problems, this is often written j so that no confusion would result with the current if it is designated by i. As described in Chapter 2, the number $\overline{z} = x - iy$ is the complex *conjugate* of z. The notation z^* is also used to designate the complex conjugate of z.

Important properties of *real* polynomials and their roots are as follows:

- 1. A polynomial of degree $n \geq 1$ has n roots.
- 2. A polynomial of odd degree has at least one real root.
- 3. If z is a complex root of a real polynomial, then the complex conjugate \bar{z} is a root also.

The polynomial $P(z) = a_n z^n + a_{n-1} z^{n-1} + \dots + a_0$ can always be written in the form

$$P(z) = (z - z_1)(z - z_2) \cdots (z - z_n)a_n$$
(3)

as the product of linear factors using the roots $z_i, i = 1, 2, \ldots, n$ of

$$P(z) = 0.$$

See also **conv** and **deconv** Harman p431 and **polyval** Harman p350.

Polynomial fit (7.1, P351) Go over Example 7.1 Harman P353.

Handout 4. Functions, Sequences, Series (Harman P 276) Review Harman Pages 276-277,282-296.

- 1. Continuous functions definition P276
- 2. Sequences Section 6.2 P 282
- 3. Infinite Series P 284-287
- 4. Geometric Series P 287
- 5. Series of functions P 288
- 6. Power Series P 290

Taylor Series (6.3, P292)

After a review of Taylor series, we can visit Euler's formula again using results in Table 6.3 P295. Consider the Taylor series

$$e^{i\theta} = 1 + i\theta + \frac{(i\theta)^2}{2!} + \frac{(i\theta)^3}{3!} + \frac{(i\theta)^4}{4!} + \cdots$$

Collecting terms using the fact that $i^2 = -1$, $i^3 = -i$, and so on yields

$$e^{i\theta} = \left(1 - \frac{(\theta)^2}{2!} + \frac{(\theta)^4}{4!} - \cdots\right) + i\left(\theta - \frac{(\theta)^3}{3!} + \frac{(\theta)^5}{5!} - \cdots\right)$$
$$= \cos\theta + i\sin\theta \tag{4}$$

Notice that in all the Taylor series examples, if the argument such as θ in the expansions for sin and cos only a few terms may be needed to satisfy the precision requirements of a problem. For example, if θ is 0.05 radians (about 3 degrees),

$$\sin(0.05) = 0.0500$$

to four decimal places. In other words, $\sin x \approx x$ when x is small.



Figure 1: Caption for FullWaveRectscan0001

EXAMPLE Series Output of a Full Wave Rectifier

When the diodes are reversed biased by the voltage V_p on the capacitor, the voltage across the resistor and capacitor are equal so that

$$i = \frac{v}{R} = C\frac{dv}{dt}$$

The differential equation

$$\frac{dv}{dt} + \frac{v}{CR}$$
 with initial conditions $v(0) = V_P$

The solution is $v(t) = V_p e^{\lambda t}$ with λ the solution to the *characteristic* equation $\lambda + 1/RC = 0$ as described in Harman p216. The decay in the voltage from the peak is computed by expanding the exponential and evaluating the first terms at t = T/2 to yield

$$V_{min} = V_p \exp^{\left(\frac{-t}{RC}\right)} \approx V_p \left(1 - \frac{T/2}{RC}\right) = V_p \left(1 - \frac{1}{2fRC}\right)$$
(5)

The figures illustrate the problem. The idea is to determine the ripple.



Figure 2.11 The ripple is the undesirable portion of the output. (a) Rectifier output; (b) dc component; (c) ripple component.







Figure 2.13 Full-wave bridge rectifier with a capacitor filter.



Figure 2: Caption for 5131FullWaveCogdell0001

The input wave in is a 120 volt rms, 60 Hertz sine wave so that $v(t) = 120\sqrt{2}\sin(2\pi 60t)$. The time constant for the decay is

$$\tau = RC = 100 \times 2000 \times 10^{-6} = 200 \ ms$$

which is long compared to the period of the sine wave at T = 1/60 = 16.67 ms. Using the values in the series yields a value of about 0.04 for the exponent at t = T/2 which indicates that the approximation to V_{min} is reasonable.



* OUR FOLVATION 5

Figure 3: Cogdell FullWave Ex2_14